

EECS20n, Solution to Mock Midterm 2, 11/17/00

1. **15 points** Write the following in Cartesian coordinates (i.e. in the form $x + jy$)

(a) **1 point** $j^3 - j^2 + j - 1 = 0$

(b) **2 points** $\sum_{k=0}^{11} e^{jk\pi/6} = \frac{1 - e^{j12\pi/6}}{1 - e^{j\pi/6}} = 0$

(c) **2 points** $(1 + j1)/(1 - j1) = j$

(d) **2 points** $\sqrt{\cos \pi/4 + j \sin \pi/4} = \sqrt{e^{j\pi/4}} = \pm e^{j\pi/8} = \pm(\cos \pi/8 + j \sin \pi/8).$

Write the following in polar coordinates (i.e. in the form $r e^{j\theta}$)

(a) **2 points** $1 + j1 = \sqrt{2}e^{j\pi/4}$

(b) **2 points** $(1 + j1)^3 = 3\sqrt{2}e^{j\pi/4}$

(c) **2 points** $[\cos \pi/4 + j \sin \pi/4]^{1/2} = \pm 1.e^{j\pi/8}$

(d) **2 points** $(1 + j1)/(1 - j1) = 1.e^{j\pi/2}$

2. **15 points** Which of the following discrete-time or continuous-time signals is periodic. Answer yes or no. If the signal is periodic, give its fundamental period and state the units. Suppose that for a discrete-time signal, n denotes **seconds**, and for a continuous-time signal, t denotes **minutes**.

(a) **2 points** $\forall n \in \text{Ints}, \quad x(n) = e^{j\sqrt{2}n}$ Periodic NO

(b) **2 points** $\forall t \in \text{Reals}, \quad x(t) = e^{j\sqrt{2}t}$ Periodic YES, Period = $2\pi/\sqrt{2}$ min.

(c) **2 points** $\forall n \in \text{Ints}, \quad x(n) = \cos 3\pi n + \sin(3\pi n + \pi/7)$ Periodic YES, Period = 2 sec.

(d) **2 points** $\forall t \in \text{Reals}, \quad x(t) = \cos 3t + |\sin 3t|$ Periodic YES, Period = $2\pi/3$ min.

(e) **2 points** $\forall n \in \text{Ints}, \quad x(n) = |\cos 3\pi n| + \sin(3\pi n + \pi/7)$ Periodic YES, Period = 2 sec.

5 points Find A, θ, ω in the following expression:

$$\begin{aligned} A \cos(\omega t + \theta) &= \cos(2\pi \times 10,000t + \frac{\pi}{4}) + \sin(2\pi \times 10,000t + \frac{\pi}{4}) \\ &= \sqrt{2} \cos(2\pi \times 10,000t) \end{aligned}$$

So, $A = \sqrt{2}, \omega = 20,000\pi, \theta = 0$.

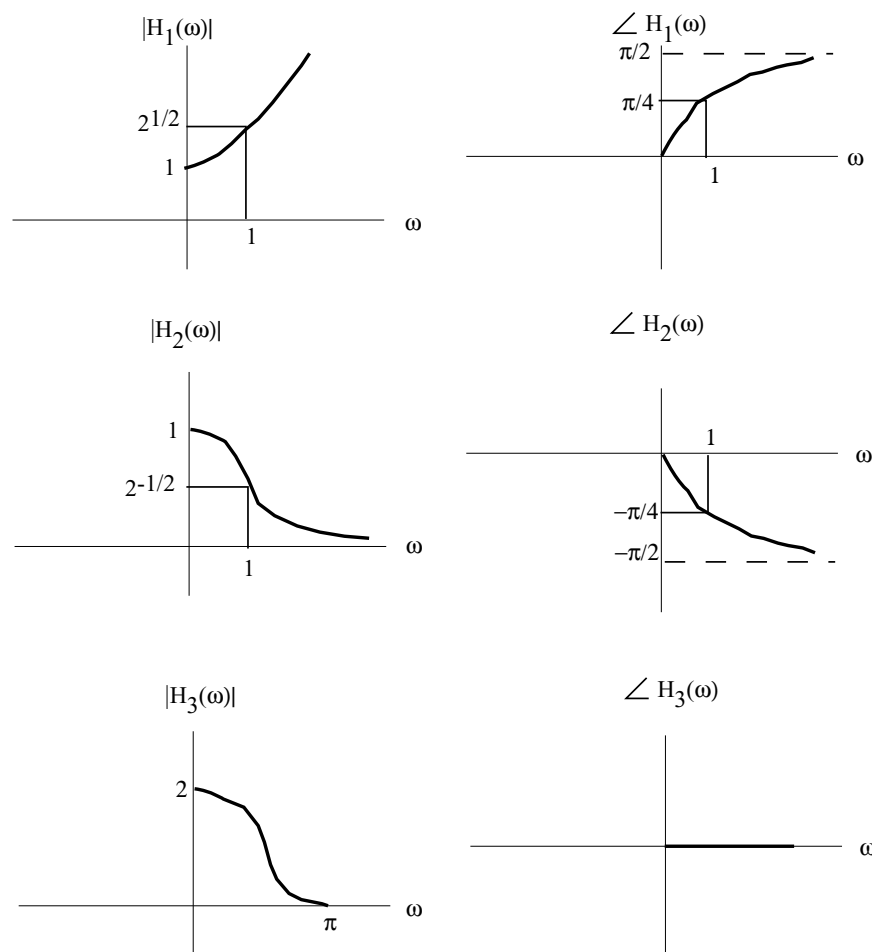


Figure 1: Plots for Problem 3

3. **15 points** On Figure 1 plot the amplitude and phase response of the following frequency responses. On your plots carefully mark the values for $\omega = 0$ and for one other non-zero value of ω .

(a) **4 points** $\forall \omega \in \text{Reals}, H_1(\omega) = 1 + j\omega$.

So, $|H_1(\omega)| = [1 + \omega^2]^{1/2}$, $\angle H_1(\omega) = \tan^{-1} \omega$. $|H_1(0)| = 1, |H_1(1)| = \sqrt{2}$,
 $\angle H_1(0) = 0, \angle H_1(1) = \pi/4$.

(b) **4 points** $\forall \omega \in \text{Reals}, H_2(\omega) = \frac{1}{1+j\omega}$. So, $|H_2(\omega)| = [1 + \omega^2]^{-1/2}$, $\angle H_2(\omega) = -\tan^{-1} \omega$. $|H_2(0)| = 1, |H_2(1)| = 1/\sqrt{2}$, $\angle H_2(0) = 0, \angle H_2(1) = -\pi/4$.

(c) **4 points** $\forall \omega \in \text{Reals}, H_3(\omega) = 1 + \cos \omega$.

So $|H_3(\omega)| = 1 + \cos \omega, \angle H_3(\omega) = 0$. $H_3(0) = 2, H_3(\pi) = 0$.

3 points Which of H_1, H_2, H_3 can be the frequency response of a discrete-time system?
 H_3 . Since it is periodic with period 2π .

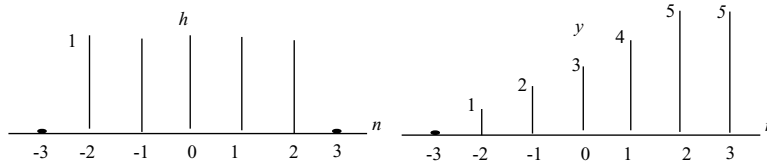


Figure 2: Impulse and step response for Problem 4

4. **15 points** A discrete-time system H has impulse response $h : \text{Ints} \rightarrow \text{Reals}$ given by

$$h(n) = \begin{cases} 1, & n = -2, -1, 0, 1, 2 \\ 0, & \text{otherwise} \end{cases}$$

- (a) **2 points** Sketch h .
- (b) **5 points** What is the step response of H , i.e. the output signal when the input signal is *step*, where $\text{step}(n) = 1, n \geq 0$, and $\text{step}(n) = 0, n < 0$? You can give your answer as a plot or as an expression.
- (c) **5 points** What is the frequency response of H ?
- (d) **3 points** What is the output signal of H for the following input signals?
- $\forall n, x(n) = \cos n$
 - $\forall n, x(n) = \cos(n + \pi/6)$
 - $\forall n, x(n) = \sin 100n$

(a,b)Figure 2 shows h and the step response y . An alternative way to calculate y is

$$\begin{aligned} \forall n \in \text{Ints}, \quad y(n) &= \sum_{m=-\infty}^{\infty} h(m)x(n-m) \\ &= x(n+2) + x(n+1) + x(n) + x(n-1) + x(n-2) \\ &= 0, n \leq -3; 1, n = -2; 2, n = -1; 3, n = 0; 4, n = 1; 5, n \geq 2. \end{aligned}$$

(c) The frequency response is

$$\begin{aligned} \hat{H}(\omega) &= \sum_{n=-\infty}^{\infty} h(n)e^{-j\omega n} \\ &= \sum_{n=-2}^2 e^{-j\omega n} \\ &= 1 + 2 \cos \omega + 2 \cos 2\omega \end{aligned}$$

(d) Use the fact that the response to a signal $x(n) = \cos(\omega n + \theta)$ is $y(n) = |\hat{H}(\omega)| \cos(\omega n + \theta + \angle \hat{H}(\omega))$.

- $\forall n, y(n) = \hat{H}(1)x(n) = [1 + 2 \cos 1 + 2 \cos 2] \cos n$
- $\forall n, y(n) = \hat{H}(1)x(n) = [1 + 2 \cos 1 + 2 \cos 2] \cos(n + \pi/6)$
- $\forall n, y(n) = \hat{H}(100)x(n) = [1 + 2 \cos 100 + 2 \cos 200] \sin 100n$

5. 15 points

(a) **5 points** Find the frequency response for the LTI systems described by these differential equations (input is x , output is y)

i. $\dot{y}(t) - 0.5y(t) = x(t)$

ii. $\ddot{y}(t) - 0.5\dot{y}(t) + 0.25y(t) = \dot{x}(t) + x(t)$

(b) **5 points** What is the response of the second system above for the input $\forall t, x(t) = e^{j(100t+\pi/4)}$?

(c) **5 points** Find the frequency response for the LTI systems described by these difference equations (input is x , output is y)

i. $y(n) - 0.5y(n-1) = x(n)$

ii. $y(n) - y(n-1) + 0.25y(n-2) = x(n) + x(n-1)$

(a) Using $y(t) = \hat{H}(\omega)e^{j\omega t}$ is the response to $x(t) = e^{j\omega t}$, we get

i. $\frac{1}{j\omega - 0.5}$

ii. $\hat{H}(\omega) = \frac{j\omega + 1}{-\omega^2 - 0.5j\omega + 0.25} = \frac{1 + j\omega}{0.25 - \omega^2 - 0.5j\omega}$

(b) The response is $\forall t, \hat{H}(100)e^{j(100t+\pi/4)}$.

(c) Using $y(n) = \hat{H}(\omega)e^{j\omega n}$ is the response to $x(n) = e^{j\omega n}$, we get

i. $\frac{1}{1 - 0.5e^{-j\omega}}$

ii. $\frac{1 + e^{-j\omega}}{1 - e^{-j\omega} + 0.25e^{-2j\omega}}$.

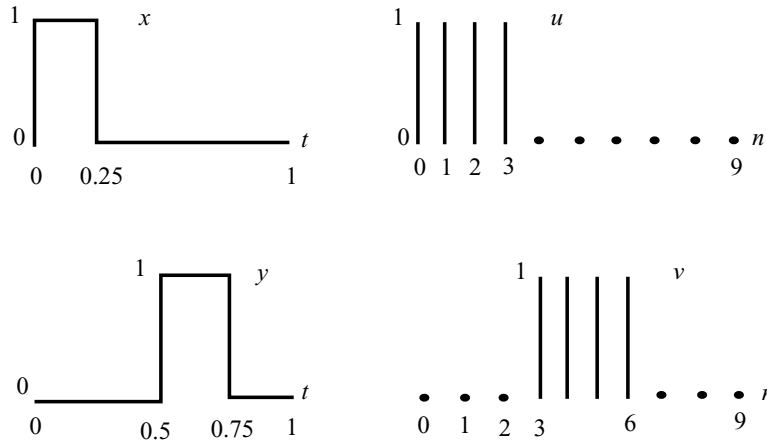


Figure 3: Periodic signals for Problem 6

6. **15 points** Figure 3 plots two continuous-time periodic signals x and y both with period 1 second, and two discrete-time signals u and v both with period 10 samples. The plots are given only for one period. Suppose the exponential Fourier Series representations of these signals are given as:

$$\forall t \in \text{Reals}, \quad x(t) = \sum_{k=-\infty}^{\infty} X_k e^{jk\omega_x t}$$

$$\forall t \in \text{Reals}, \quad y(t) = \sum_{k=-\infty}^{\infty} Y_k e^{jk\omega_y t}$$

$$\forall n \in \text{Ints}, \quad u(n) = \sum_{k=0}^9 U_k e^{jk\omega_u n}$$

$$\forall n \in \text{Ints}, \quad v(n) = \sum_{k=0}^9 V_k e^{jk\omega_v n}$$

- (a) **3 points** Give the values of $\omega_x = 2\pi$ rad/sec, $\omega_y = 2\pi$ rad/sec, $\omega_u = \pi/5$ rad/sample, $\omega_v = \pi/5$ rad/sample.
- (b) **4 points** Calculate the values of the coefficients $X_0 = 0.25$, $Y_0 = 0.25$, $U_0 = 0.4$, $V_0 = 0.4$.
These are just the average values of the signal over one period.
- (c) **2 points** Suppose the signals x is measured in Volts. What is the unit of X_0 ? Volts.
- (d) **4 points** Calculate the values of the coefficients $X_1 = \int_{t=0}^{0.25} e^{-j2\pi t} dt = \frac{1-j}{2\pi}$
 $Y_1 = \int_{t=0.5}^{0.75} e^{-j2\pi t} dt = \frac{-1+j}{2\pi}$
 $U_1 = \frac{1}{10} \sum_{n=0}^9 x(n) e^{-j\pi/5 n} = \frac{1}{10} \sum_{n=0}^3 e^{-j\pi/5 n} = \frac{1-e^{-j4\pi/5}}{10(1-e^{-j\pi/5})}$
 $V_1 = \frac{1}{10} \sum_{n=3}^6 e^{-j\pi/5 n} = e^{-j3\pi/5} \frac{1-e^{-j4\pi/5}}{10(1-e^{-j\pi/5})}$
- (e) **3 points** Express y as a delayed version of x and v as a delayed version of u .
 $\forall t, y(t) = x(t - 0.5), \forall n, v(n) = x(n - 3)$.

- (f) **4 points** Express the FS coefficients $\{Y_k\}$ in terms of $\{X_k\}$ and $\{V_k\}$ in terms of $\{U_k\}$.
 $\forall k, \quad Y_k = X_k e^{-jk\pi}, \forall k, \quad V_k = U_k e^{-jk3\pi/5}.$

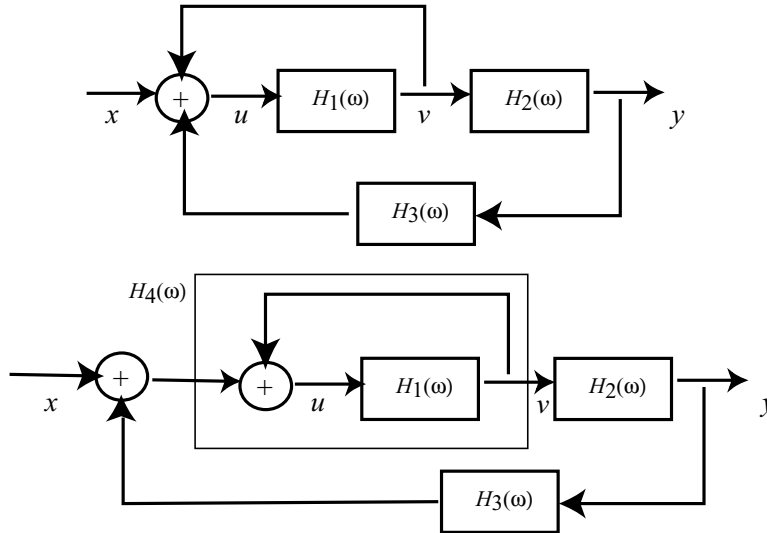


Figure 4: Feedback systems for Problem 7

7. **15 points** Figure 4 shows two feedback systems. In these figures, $H_k(\omega)$, $k = 1, 2, 3$ is the frequency response of the three LTI systems.

(a) **9 points** Calculate the closed-loop frequency response $H(\omega)$ of the first feedback system in terms of the H_k . Hint: Use the fact that the two systems are the same.

By the hint, $H_4 = H_1/(1 - H_1)$. So,

$$H = \frac{H_4 H_2}{1 - H_4 H_2 H_3} = \frac{H_1 H_2}{1 - H_1 - H_1 H_2 H_3} \quad (1)$$

(b) **6 points** Suppose $H_k(\omega) = 1/(1 + j2\omega)$ for all $k = 1, 2, 3$. Calculate $H(0)$, $H(1)$ and $\lim_{\omega \rightarrow \infty} H(\omega)$.

We have $H_k(0) = 1$, $H_k(1) = 1/(1 + j2)$, $\lim_{\omega \rightarrow \infty} H_k(\omega) = 0$. Substitution in (1) gives,

$$H(0) = -1, H(1) = \frac{\frac{1}{(1+2j)^2}}{1 - \frac{1}{1+2j} - \frac{1}{(1+2j)^3}} = \frac{1 + 2j}{(1 + 2j)^3 - (1 + 2j)^2 - 1}$$

$$\lim_{\omega \rightarrow \infty} H(\omega) = 0.$$

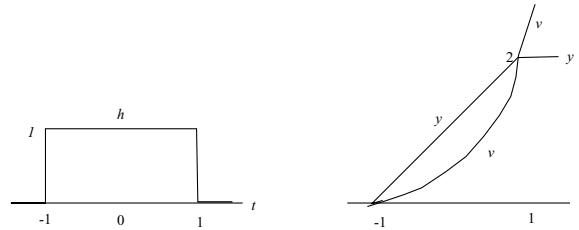


Figure 5: Impulse response for Problem 8

8. **15 points** A continuous-time LTI system has the impulse response

$$\forall t \in \text{Reals}, \quad h(t) = \begin{cases} 1, & |t| < 1 \\ 0, & \text{otherwise} \end{cases}$$

(a) **3 points** Sketch the impulse response, and mark carefully the relevant points on your plot.

This is shown in Figure 5

(b) **3 points** Is this system causal? Answer yes or no. NOT CAUSAL.

(c) **3 points** What is the step response of this system, i.e. the response to $\text{step}(t) = 1, t \geq 0$ and $= 0, t < 0$?

The step response is the integral of the impulse response,

$$y(t) = \int_{s=-\infty}^t h(s) ds = \begin{cases} 0, & t \leq -1 \\ t + 1, & -1 \leq t \leq 1 \\ 2, & t \geq 1 \end{cases}$$

A sketch of y is in the figure.

(d) **3 points** What is the ramp response of this system, i.e. the response to $\text{ramp}(t) = t, t \geq 0$, and $= 0, t < 0$?

The ramp response v is the integral of the step response,

$$v(t) = \int_{s=-\infty}^t y(s) ds = \begin{cases} 0, & t \leq -1 \\ \frac{1}{2}(t+1)^2, & -1 \leq t \leq 1, \\ 2 + 2(t-1), & t \geq 1 \end{cases}$$

(e) **3 points** What is the response of this system to the input signal impulsetrain , where

$$\forall t \in \text{Reals}, \quad \text{impulsetrain}(t) = \sum_{k=-\infty}^{\infty} \delta(t - 2k).$$

The response is

$$\begin{aligned} \forall t, \quad w(t) &= h * \text{impulsetrain}(t) \\ &= \int_{\tau=-\infty}^{\infty} h(t - \tau) \text{impulsetrain}(\tau) d\tau \end{aligned}$$

$$\begin{aligned} &= \sum_{k=-\infty}^{\infty} \int_{\tau=-\infty}^{\infty} h(t-\tau)\delta(\tau-2k) \\ &= \sum_{k=-\infty}^{\infty} h(t-2k) = 1. \end{aligned}$$