UNIVERSITY OF CALIFORNIA

College of Engineering Department of Electrical Engineering and Computer Sciences

Professor Tse Spring 2005

EECS 20n — Final Exam Solutions

[5 pts.] Problem 1

$$y(n) = \frac{1}{4}y(n-1) + x(n) + \frac{1}{2}x(n-1)$$

This is a linear constant coefficient system that is initially at rest \rightarrow LTI .

[15 pts.] Question 2

(a) (10 pts.)
$$q(n) = \frac{1}{2}e^{i\frac{2\pi}{3}n} + \frac{1}{2}e^{-i\frac{2\pi}{3}n} + e^{i\pi n}$$

$$period 3 \qquad period 2$$
Overall period = 6
$$\omega_0 = \frac{2\pi}{6}$$

$$X_2 = X_{-2} = \frac{1}{2}$$

$$X_3 = 1 \qquad \text{for } -2 \le k \le 3$$

$$X_k = 0 \quad \text{otherwise}$$

(b) (5 pts.)
$$v(t) = \frac{\cos(\pi t) + \cos(t)}{\text{period 2}} + \frac{\cos(\pi t)}{\text{period 2}\pi}$$

Overall period $p = 2m = 2\pi n$ where m, n are integers.

No such m, n exists because 2 is rational and 2π is irrational, \therefore not periodic.

[10 pts.] Question 3

(a) (5 pts.) Yes, the system F can be linear. x(t) $cos(\omega_0 t)$ y(t)

(b) (5 pts.) The system F cannot be time-invariant, because new frequencies have been created that did not exist in $X(\omega)$.

[15 pts.] Question 4

(a) (7 pts.) Reading the signals off the diagram, we have:

$$y(n) = -5y(n-1) - y(n-1) + x(n-1) + x(n)$$

Written alternatively, the LCCDE describing the system above is:

$$y(n) + 5y(n-1) + y(n-1) = x(n) + x(n-1)$$

(b) (8 pts.) We note that if $s_i(n)$ is the output of a delay block, then the input to the delay block must be $s_i(n+1)$, as shown below:

$$s_i(n+1) \longrightarrow D \longrightarrow s_i(n)$$
 $i = 1, 2$

Accordingly, we can label the original delay-adder-gain block diagram with $s_1(n+1)$ and $s_2(n+1)$.

We can now read off the diagram the expressions for $s_1(n+1)$, $s_2(n+1)$, and y(n).

$$s_{1}(n+1) = s_{2}(n) - 5y(n) + x(n) \} \Rightarrow s_{1}(n+1) = -5s_{1}(n) + s_{2}(n) - 4x(n)$$

$$y(n) = s_{1}(n) + x(n)$$

$$s_{2}(n+1) = -y(n)$$

$$\Rightarrow s_{2}(n+1) = -s_{1}(n) - x(n)$$

Hence, the state-space equations are:

$$\begin{bmatrix} s_1(n+1) \\ s_2(n+1) \end{bmatrix} = \begin{bmatrix} -5 & 1 \\ -1 & 0 \end{bmatrix} \begin{bmatrix} s_1(n) \\ s_2(n) \end{bmatrix} + \begin{bmatrix} -4 \\ -1 \end{bmatrix} x(n)$$

$$y(n) = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} s_1(n) \\ s_2(n) \end{bmatrix} + 1 \cdot x(n)$$

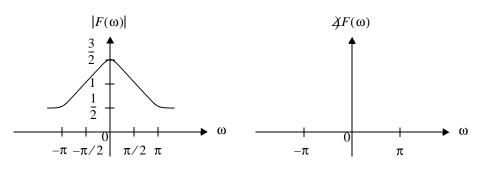
$$c^T \begin{bmatrix} s_1(n) \\ s_2(n) \end{bmatrix} + 1 \cdot x(n)$$

[25 pts.] Question 5

(a) (7 pts.)
$$F(\omega) = \sum_{n} f(n)e^{-i\omega n} = \frac{1}{4}e^{i\omega} + 1 + \frac{1}{4}e^{-i\omega} = 1 + \frac{1}{2}\cos\omega$$

$$F(\omega) \in \mathbb{R} \atop F(\omega) > 0$$
 $\Rightarrow F(\omega) = |F(\omega)| = 1 + \frac{1}{2}\cos\omega$

$$\cancel{\angle}F(\omega) = 0 \text{ for a positive real quantity}$$

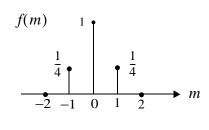


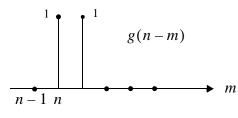
Note: $F(\omega)$, $|F(\omega)|$, and $\angle F(\omega)$ are periodic and repeat outside the $(-\pi, \pi)$ interval.

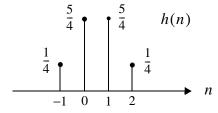
(b) (8 pts.) h = f * g for a cascade interconnection.

$$h(n) = \sum_{m} f(m)g(n-m)$$

$$h(n) = \begin{cases} \frac{1}{4} & n = -1, 2 \\ \frac{5}{4} & n = 0, 1 \\ 0 & \text{elsewhere} \end{cases}$$







Sanity Check:

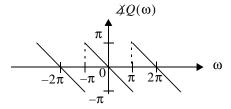
- f(n) has a region of support of length N = 3.
- g(n) has a region of support of length M = 2.
- h(n) has a region of support of length N + M 1 = 4.

(c) (10 pts.) Recognize that g(n) = f(n-1) of Part (a). Hence,

$$Q(\omega) = F(\omega)e^{-i\omega} = \left(1 + \frac{1}{2}\cos\omega\right)e^{-i\omega} \Rightarrow$$

$$|Q(\omega)| = |F(\omega)| = 1 + \frac{1}{2}\cos\omega \qquad \cancel{\angle} Q(\omega) = -\omega \qquad -\pi < \omega < \pi$$
Periodic with period 2π

We plot only the phase here:



$$r(n) = \cos\left(\frac{7\pi n}{3} + \frac{\pi}{3}\right) = \cos\left(\frac{\pi n}{3} + \frac{\pi}{3}\right)$$
$$\omega_0 = \frac{\pi}{3}; \ \theta = \frac{\pi}{3}$$

For an LTI system

$$\cos(\omega_0 n + \theta) \longrightarrow \boxed{Q(\omega) \\ LTI} \qquad |Q(\omega_0)| \cos(\omega_0 n + \theta + \cancel{\chi} Q(\omega_0))$$

This can be verified by recasting $\cos(\omega_0 n + \theta)$ in terms of complex exponentials and simplifying expressions after invoking the eigenfunction property of complex exponentials. Hence, the output v(n) is:

$$v(n) = \left| Q\left(\frac{\pi}{3}\right) \right| \cos\left(\frac{\pi n}{3} + \frac{\pi}{3} + \cancel{\mathcal{L}}Q\left(\frac{\pi}{3}\right)\right) = \frac{5}{4}\cos\left(\frac{\pi n}{3} + \frac{\pi}{3} - \frac{\pi}{3}\right)$$
$$v(n) = \frac{5}{4}\cos\left(\frac{\pi n}{3}\right)$$

[15 pts.] Question 6

(a) (4 pts.) The unit of ω_0 is **radians/second**.

The unit of X_k is **volts**.

(b) (6 pts.) The unit of ω is **radians/second**.

The unit of $d\omega$ is radians/second.

The unit of $X(\omega)$ is: $\frac{X(\omega)d\omega}{2\pi}$ has unit of volt $\Rightarrow X(\omega)$ has unit of $\frac{\text{volt} \cdot \text{rad}}{\text{rad/sec}} = \text{volt} \cdot \text{sec}$.

No, X_k and $X(\omega)$ do not have the same unit; X_k has the same unit as $\frac{X(\omega)d\omega}{2\pi}$.

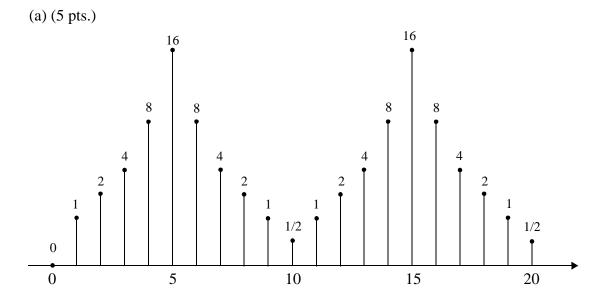
(c) (5 pts.) The unit of Ω is **radians/sample**.

The unit of $d\Omega$ is **radians/sample**.

The unit of $Y(\Omega)$ is: $\frac{Y(\Omega)d\Omega}{2\pi}$ has unit of volt $\Rightarrow Y(\Omega)$ has unit of $\frac{\text{volt} \cdot \text{radians}}{\text{radians/sample}} = \text{volt} \cdot \text{sample}$.

No, the units of ω and Ω are not the same. ω is the frequency of a CT signal \Rightarrow has unit of radians/sec. Ω is the frequency of a DT signal \Rightarrow has unit of radians/sample. Again here, 2π has unit of radians.

[20 pts.] Question 7



(b) (9 pts.) No, overall state
$$S = \{1, 2\} \times \text{Reals}$$

state = $(\text{mode}(n), r(n))$

mode(n)	r(n)	mode(n+1)	r(n+1)	output $y(n)$
1	$ r(n) \le 10$	1	2r(n) + x(n)	r(n)
1	r(n) > 10	2	$\frac{1}{2}r(n) + x(n)$	r(n)
2	r(n) < 1	1	2r(n) + x(n)	r(n)
2	$ r(n) \ge 1$	2	$\frac{1}{2}r(n) + x(n)$	r(n)

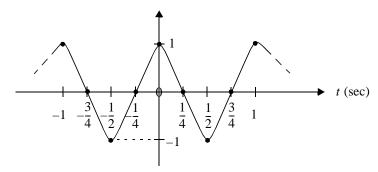
(c) (6 pts.) state(n) = (mode(n), time in mode 2,
$$r(n)$$
)
$$S = \{1, 2\} \times \{0, 1, 2\} \times \text{Reals}$$

[10 pts.] Question 8

(a) (5 pts.) We sample four times every second, at equally-spaced points in time:

$$T - 0.25 \text{ sec } \Rightarrow f_s = 4 \text{ Hz}$$
 $(\omega_s = 8\pi \text{ rad/sec})$

where s is the sampling frequency.



(b) (5 pts.)
$$f_s - f_2$$
 will appear as $f_1 \Rightarrow 4 = f_2 = f_1 = 1 \Rightarrow f_2 = 3$ Hz.

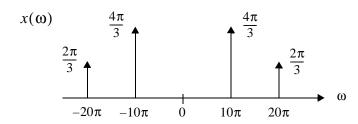
Verify:
$$y(t) = \cos(6\pi t)$$
 evaluated at $t = 0, \frac{1}{4}, \frac{1}{2}, \frac{3}{4}$.

Second points:

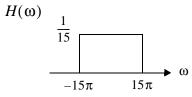
$$y(0) = \cos(0) = 1 = x(0); \ y\left(\frac{1}{4}\right) = \cos\left(\frac{6\pi}{4}\right) = \cos\left(\frac{3\pi}{2}\right) = 0 = x\left(\frac{1}{4}\right)$$
$$y\left(\frac{1}{2}\right) = \cos\left(\frac{6\pi}{2}\right) = \cos(3\pi) = -1 = x\left(\frac{1}{2}\right); \text{ and } y\left(\frac{3}{4}\right) = \cos\left(\frac{6\pi \cdot 3}{4}\right) = \cos\left(\frac{9\pi}{2}\right) = 0$$
$$= x\left(\frac{3}{4}\right)$$

[15 pts.] Question 9

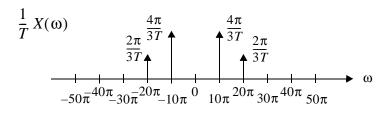
(a) (5 pts.)
$$\omega_0 = 10\pi \text{ rad/sec} \implies x_1 = \frac{2}{3} \stackrel{F}{\iff} e^{i10\pi t}; \quad x_{-1} = \frac{2}{3} \stackrel{F}{\iff} e^{-i10\pi t}$$
$$x_2 = \frac{1}{3} \stackrel{F}{\iff} e^{i20\pi t}; \quad x_{-2} = \frac{1}{3} \stackrel{F}{\iff} e^{-i20\pi t}$$
$$\implies x(t) = \frac{2}{3} (e^{i10\pi t} + e^{-i10\pi t}) + \frac{1}{3} (e^{i20\pi t} + e^{-i20\pi t}) \implies$$
$$x(t) = \frac{4}{3} \cos(10\pi t) + \frac{2}{3} \cos(20\pi t)$$



(b) (10 pts.)
$$f_s = 15 \text{ Hz } (\omega_s = 30\pi \text{ rad/sec}) \Rightarrow X_p(\omega) = \frac{1}{T} \sum_k X(\omega - k\omega_s) = 15 \sum_k X(\omega - 30k\pi)$$



Because of the filter $H(\omega)$, all frequency content of $X_p(\omega)$ outside of the $(-15\pi, 15\pi)$ band will be suppressed. \Rightarrow We need only look at $\frac{1}{T}X(\omega)$, $\frac{1}{T}X(\omega-30\pi)$, and $\frac{1}{T}X(\omega+30\pi)$:



$$\frac{1}{T} X(\omega - 30\pi)$$

$$\frac{2\pi}{3T} \stackrel{4\pi}{4} \stackrel{\pi}{3T}$$

$$\frac{2\pi}{3T} \stackrel{4\pi}{4} \stackrel{\pi}{3T}$$

$$\frac{2\pi}{3T} \stackrel{\pi}{4} \stackrel{\pi}{3T}$$

$$-50\pi^{40\pi} - 30\pi^{-20\pi} - 10\pi^{0}$$

$$10\pi^{20\pi} 30\pi^{40\pi} 50\pi$$

$$0$$

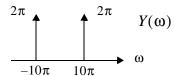
$$\frac{1}{T}X(\omega + 30\pi) \xrightarrow{4\pi \over 3T} \xrightarrow{4\pi \over 3T}$$

$$\frac{2\pi}{3T} \xrightarrow{4\pi \over 3T}$$

$$-50\pi^{-40\pi} - 30\pi^{-20\pi} - 10\pi^{-0}$$

$$10\pi^{-20\pi} 30\pi^{-40\pi} 50\pi$$

The above three components [scaled spectral replica of $X(\omega)$] are added and only the frequency content in the $(-15\pi, 15\pi)$ band is retained (and scaled). Hence, the spectrum of y(t) is:



 $\Rightarrow y(t) = 2\cos(10\pi t)$ (where 20π frequency components are aliased down to 10π).